## Types of Filters

We saw three types of filters, viz. MA, AR, and ARMA. These do not include the entire class of filters but cover an important class of filters and are very useful.

For linear systems where output $y(n)$ is related to the input $x(n)$ by the difference equation:

$$
y(n)+\sum_{k=1}^{p} g_{k} y(n-k)=\sum_{k=0}^{q} \mathrm{f}_{\mathrm{k}} x(n-k)
$$

We considered three cases:

## 1. Moving Average Filters (MA)

In this case, the linear filter, is an all-zero filter and the difference equation for their input-output relationship is

$$
y(k)=x_{k}+f_{0} x_{k-1}+f_{1} x_{k-2}+\ldots
$$

Note: The strength of these filters may not be 1


## 2. Autoregressive Filters (AR)

In this case, the linear filter, is an all-zero filter and the difference equation for their input-output relationship is

$$
y(k)=x_{k}+g_{0} y_{k-1}+g_{1} y_{k-2}+\ldots
$$



## AR Filters are linear shift invariant systems

Refer to previous AR filter figure.
Using the figure we can calculate the expressions for $\mathrm{y}_{0}, \mathrm{y}_{1} \ldots$

$$
\begin{aligned}
& y_{0}=x_{0} \\
& y_{1}=x_{1}+g_{0} x_{0} \\
& y_{2}=x_{2}+g_{0} x_{1}+\left(g_{0}{ }^{2}+g_{1}\right) x_{0} \\
& y_{3}=x_{3}+g_{0} x_{2}+\left(g_{0}{ }^{2}+g_{1}\right) x_{1}+\left(g_{0}{ }^{3}+2 g_{0} g_{1}+g_{2}\right) x_{0}
\end{aligned}
$$

This can be seen as

$$
\begin{aligned}
& Y=X * F \\
& F=\left[\begin{array}{llll}
1 & g_{0} & \left(g_{0}^{2}+g_{1}\right) & \left(g_{0}^{3}+2 g_{0} g_{1}+g_{2}\right) \ldots
\end{array}\right]
\end{aligned}
$$

By looking at the difference equation of AR filters we observe that $y_{k}$ is described in terms of $k$ and does not refer to any particular value of $y$ or $x$. Hence we can say that AR filters are shift invariant

To show that AR is linear, we have to show that


Let us superimpose the blue and red AR systems as in the figure below and finally add up the two results in $\Sigma_{2}$. We can look at the AR system as containing the summation element $\Sigma_{1}$, the delay elements $z^{-1}$ and the multiplying elements $g_{k}$.


Let us denote, for convenience, $\Sigma_{2}$ as a black filled circle.

Add $\left(\Sigma_{2}\right)$ before the addition $\left(\Sigma_{1}\right)$ or after, it doesn't make a difference...


Add $\left(\Sigma_{2}\right)$ before the multiplication $\left(\mathrm{g}_{\mathrm{k}}\right)$ or after, it doesn't make a difference...


Now, for each of the three $\Sigma_{2}$ we have the following:
Next three figures for the $\Sigma_{2}$ before $g_{2}$
Among the next three, the $2^{\text {nd }}$ and $3^{\text {rd }}$ for the $\Sigma_{2}$ before $g_{1}$
Among the next three, the $3^{\text {rd }}$ for the $\Sigma_{2}$ before $g_{0}$
At each step we combine the $\Sigma_{2}$ 's between the delay elements
Add $\left(\Sigma_{2}\right)$ before the delay $\left(\mathrm{z}^{-1}\right)$ or after, it doesn't make a difference...



Add $\left(\Sigma_{1}\right)$ before the addition $\left(\Sigma_{2}\right)$ or after, it doesn't make a difference...


Another way of showing AR filters as linear is by induction.
$\mathrm{y}_{0}=\mathrm{x}_{0}$ forms the basis of the induction hypothesis
Let us assume $\mathrm{y}_{\mathrm{k}-1}, \mathrm{y}_{\mathrm{k}-2}, \ldots$ to be linear
$y_{k}$ is defined in terms of $x_{k}$ and $y_{k-1}, y_{k-2}, \ldots$ in the AR filter difference equation.
Since $\mathrm{x}_{\mathrm{k}}$ is just added and the y terms are multiplied by constants, we can say that $\mathrm{y}_{\mathrm{k}}$ is also linear.
Hence AR filters are linear systems.

## 3. Autoregressive, Moving Average Filters (ARMA)

In this case, the linear filter,

$$
\frac{H(z)}{G(z)}=H(z) \cdot \frac{1}{G(z)}
$$

is an pole-zero filter which has both finite poles and zeroes.
ARMA filters are basically cascaded MA and AR filters.
There are 2 types ${ }^{i}$ of ARMA filters:


Type 2 filters have an advantage over Type 1 in that only one set of latches is required for their implementation.

When constructing an ARMA filter, the AR filter may be unstable. As long as the poles of the AR filter match the zeros of the MA filter, the resulting ARMA filter is stable. However, sometimes we may not be able to match the poles and the zeros perfectly. Some of the reasons are:

1. On computers, due to precision / truncation errors
2. Incapability of specifying the physical media (plant errors)

Given two vectors $y$ and $x$, if we wanted to fit them together we would scale one of them by a scalar a. Using mathematical symbols we would write it as:

$$
\begin{aligned}
& f(a)=\|y-a x\|_{2}^{2} \\
& f(a)=(y-a x)^{T}(y-a x) \\
& f(a)=y^{T} y-2 a y^{T} x+a^{2} x^{T} x
\end{aligned}
$$

This is a minimization problem where a has to be varied to minimize $f$

$$
\begin{aligned}
& \frac{d f(a)}{d a}=2 a x^{T} x-2 y^{T} x=0 \\
& \therefore a=\frac{y^{T} x}{x^{T} x}
\end{aligned}
$$

$f(a)$ is a parabola which can have a minimum only if the coefficient of the second degree term is greater than zero viz. $x^{T} x$. This is true for an upward facing parabola as this term dominates every other terms.

For a parabola $a x^{2}+b x+c$ the minima is at $-b / 2 a$. This is obvious because the intersections with the x -axis are at $-b / a$ and $b / a$ and the parabola is symmetric ${ }^{\mathrm{ii}}$.

## Deconvolution Revisited

As we have seen before an LTI system can be represented as

$$
Y=A X
$$

Further, deconvolution is the process of finding the kernel/input given the output and the input/kernel. Hence we can view deconvolution as matrix inversion where in we need to find $X$ given $Y$ and $A$ by finding $A^{-1}$ and pre-multiplying the above equation.

A generalization of the above problem can be stated as
In most practical situations, $X$ does not span the space of $Y$ and hence there is no exact solution to the above equation. Thus we formulate the minimization problem as:

Given $Y$ and $A$, we have to get $X$ such that

$$
f(X)=\|Y-A X\|_{2} \text { is minimum. }
$$

Where $\|U\|_{2}^{2}=\sum U_{i}^{2}=U^{T} U=\langle U, U\rangle \quad\left(\mathrm{L}_{2}\right.$ Norm $)$

This problem is very similar to the problem we encountered earlier. $Y, A$ and $X$ in the above equation correspond to $y, x$, and $a$ in the earlier problem. There $a$ was a scalar here $X$ is a vector. We can see this as multiplying each multiplying each column of $A$ (a vector) with each row of $X$ (a scalar).

Thus it may seem that we have reduced this problem into multiple instances of the previous problem. However, this is not so because the columns of $A$ do not form an orthogonal basis.

The problem can be seen when we consider the following figure in which $Y$ is a vector in the plane of $a_{1}$ and $a_{2}(A$ spans $Y)$. Let us try and adopt the method of the previous problem here. We project $Y$ onto $a_{1}$ and project whatever is remaining onto $a_{2}$. We see that we are still left with some amount of $Y$ and we have to repeat the same procedure again (and again...). Although this procedure converges, it takes a lot of time.

The figure below shows the first few steps in the projection and re-projection of $Y$ along $a_{1}$ and $a_{2}$


## Least Square Matrix Inversion

$$
\begin{aligned}
& f(X)=\|Y-A X\|_{2}^{2} \\
& f(X)=Y-\left[\begin{array}{cccc}
a_{1} & a_{2} & \cdots & a_{n} \\
\vdots & \vdots & \cdots & \vdots \\
\vdots & \vdots & \cdots & \vdots \\
\vdots & \vdots & \cdots & \vdots
\end{array}\right] \cdot\left[\begin{array}{c}
x_{1} \\
\vdots \\
\vdots \\
x_{n}
\end{array}\right] \\
& f(X)=(Y-A X)^{T}(Y-A X) \\
& f(X)=\left(Y^{T}-X^{T} A^{T}\right)(Y-A X) \\
& f(X)=X^{T}\left(A^{T} A\right) X-\left(Y^{T} A X\right)-\left(Y^{T} A X\right)^{T}+Y^{T} Y \\
& f(x)=X^{T} \underbrace{\left(A^{T} A\right)}_{P} X-\underbrace{\left(2 Y^{T} A\right)}_{Q^{T}} X+\underbrace{Y^{T} Y}_{R}
\end{aligned}
$$

We write the above equation as
$f(X)=X^{T} P X+Q^{T} X+R$
where $P=A^{T} A$
$Q=-2 A^{T} Y$
$R=Y^{T} Y$
$f(X)$ is a field representing an n -dimensional paraboloid ${ }^{\text {iii }}$.
The above equation will have a minima only if $\forall x X^{T} P X>0$
P is the positive definite, written as $P \succ 0$
The above equation using summations:

$$
f(X)=\sum_{i} \sum_{j} P_{i j} X_{i} X_{j}+\sum_{i} Q_{i}
$$

We take the partial derivative with respect to $\mathrm{x}_{\mathrm{i}}$ in order to minimize $f(X)$
$\frac{\partial f}{\partial x_{i}}=\sum_{j \neq i}\left(P_{i j}+P_{j i}\right) X_{j}+2 P_{i i} X_{i}+Q_{i}$
$\frac{\partial f}{\partial x_{i}}=\sum_{j}\left(P_{i j}+P_{j i}\right) X_{j}+Q_{i}$
$\nabla f=\left[\begin{array}{c}\partial f / \partial x_{1} \\ \partial f / \partial x_{2} \\ \vdots \\ \partial f / \partial x_{n}\end{array}\right]=\left[\begin{array}{c}\sum_{j}\left(P_{1 j}+P_{j 1}\right) X_{j}+Q_{1} \\ \sum_{j}\left(P_{1 j}+P_{j 1}\right) X_{j}+Q_{1} \\ \vdots \\ \sum_{j}\left(P_{n j}+P_{j n}\right) X_{j}+Q_{n}\end{array}\right]$
$\nabla f=\left[\begin{array}{c}\sum_{j} P_{1 j} X_{j} \\ \sum_{j} P_{2 j} X_{j} \\ \vdots \\ \sum_{j} P_{n j} X_{j}\end{array}\right]+\left[\begin{array}{c}\sum_{j} P_{j 1} X_{j} \\ \sum_{j} P_{j 1} X_{j} \\ \vdots \\ \sum_{j} P_{j n} X_{j}\end{array}\right]+\left[\begin{array}{c}Q_{1} \\ Q_{2} \\ \vdots \\ Q_{n}\end{array}\right]$
$\nabla f=\left(\left[\begin{array}{cccc}P_{11} & P_{12} & \cdots & P_{1 n} \\ P_{21} & P_{22} & \cdots & P_{2 n} \\ \vdots & \vdots & & \vdots \\ P_{n 1} & P_{n 2} & \cdots & P_{n n}\end{array}\right]+\left[\begin{array}{cccc}P_{11} & P_{21} & \cdots & P_{n 1} \\ P_{1} & P_{22} & \cdots & P_{2 n} \\ \vdots & \vdots & & \vdots \\ P_{n 1} & P_{n 2} & \cdots & P_{n n}\end{array}\right]\right)\left[\begin{array}{c}X_{1} \\ X_{2} \\ \vdots \\ X_{n}\end{array}\right]+\left[\begin{array}{c}Q_{1} \\ Q_{2} \\ \vdots \\ Q_{n}\end{array}\right]$
$\nabla f=\left(P+P^{T}\right) X+Q$
The minima of $f(X)$ can be found by solving $\nabla f=\overrightarrow{0}$.
$\left(P+P^{T}\right) X+Q=\overrightarrow{0}$
$\therefore\left(P+P^{T}\right) X=-Q$
$X=-\left(P+P^{T}\right)^{-1} Q$
This is very similar to the solution $-b / 2 a$ if we put Q as $b$ and $P$ as $a$ (Assuming P is symmetric which it is in most practical cases).

Substituting for $P$ and $Q$
$X=-2\left(A^{T} A+A^{T} A\right)^{-1} A^{T} Y$
$X=\left(A^{T} A\right)^{-1} A^{T} Y$
$X=A^{\dagger} Y$
$\mathrm{A}^{\dagger}=\left(\mathrm{A}^{T} \mathrm{~A}\right)^{-1}$ is called the pseudo inverse ${ }^{\mathrm{iv}}$.

In the above equation $A^{T}$ is the projection matrix. It projects $Y$ onto the basis vectors of $A$. Finally ( $\left.A^{T} A\right)^{-1}$ converts the projections into linear combinands ${ }^{v}$.


In the above figure, the red vectors are the basis, the blue vectors are the projections of Y and the green vectors are the linear combinands.

Let us review what we have done. Given the vectors $Y$ and transformation we had to find $X$ such that $Y-A X$ was minimum. Usually Y has far more components than X . We have to tweak only a few parameters of the input vector X . For example,


Here $X$ is a 2-dimensional vector and $Y$ is a 4-dimensional vector. $A$ converts $X$ from its coordinate system into the coordinate system of $Y$, thus making $Y$ and $A X$ comparable. AX is the linear combination of the bases of $A$ weighted by the components of $X . A X$ still lies in its plane of $X$ while $Y$ is outside that plane. Hence we cannot directly equate them. Instead we drop perpendiculars to the space of $A$ and equate them. Dropping perpendiculars is done by premultiplying them with $A^{T}$

Hence, $\quad A^{T} A X=A^{T} Y$
And

$$
X=\left(A^{T} A\right)^{-1} A^{T} Y
$$

Note that this looks very similar to $a=\frac{y^{T} x}{x^{T} x}$ which we had found earlier.

Udayan's comments and corrections -
${ }^{i}$ Because of commutativity, type I and type II systems are equivalent. Thus, these are not two types of ARMA filters. Rather they are two ways of implementing them.
${ }^{\text {ii }}$ The parabola we are talking about is completely positive, and has no roots. But the $-b / 2 a$ formula holds for complex-root parabolas just like it does for real root parabolas.
iii "Quadratic surface"
${ }^{\text {iv }} \mathrm{A}^{\dagger}=\left(\mathrm{A}^{\mathrm{T}} \mathrm{A}\right)^{-1} \mathrm{~A}^{\mathrm{T}}$
${ }^{v}$ "Combinants"!!

